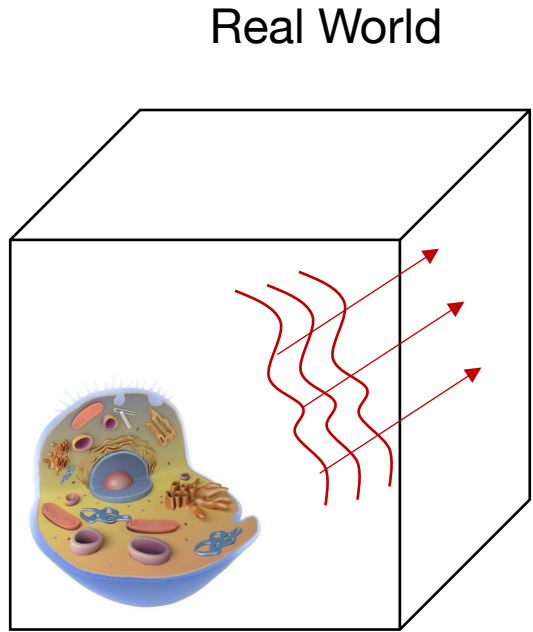


Lecture 4: Mathematical preliminaries for discrete functions

Machine Learning and Imaging

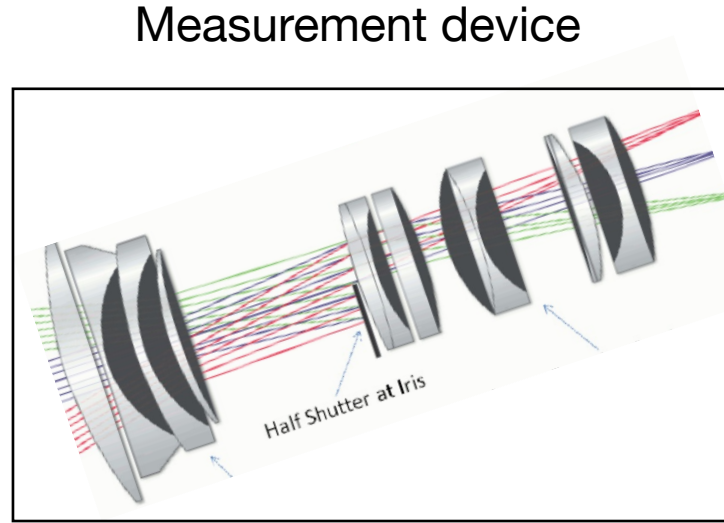
BME 548L
Roarke Horstmeyer

ML+Imaging pipeline introduction



Continuous complex fields

(last class)

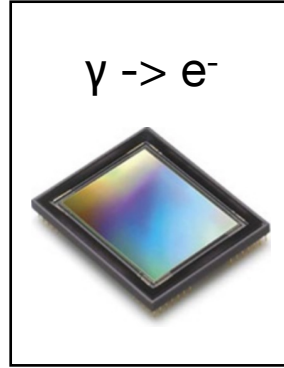


Black box transformations

- Convolution
- Fourier Transform

(last class)

Digitization

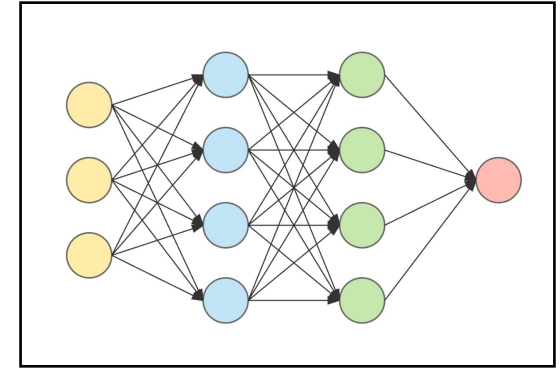


Sampling Theorem

Discrete math & Linear algebra

(this class)

Machine Learning



Optimization

Linear classification

Logistic classifier

Neural networks

Convolutional NN's

(next few weeks)



Month 2

Month 3

Review - continuous Fourier transforms – for 2D images

Decomposition of a signal into elementary functions of form, $\exp(-2\pi i(f_x x + f_y y))$:

$$\mathcal{F}\{U(x, y)\} = \hat{U}(f_x, f_y) = \iint_{-\infty}^{\infty} U(x, y) \exp(-2\pi i(f_x x + f_y y)) dx dy$$

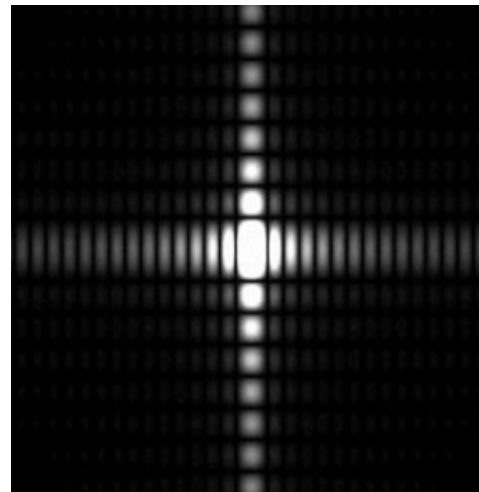
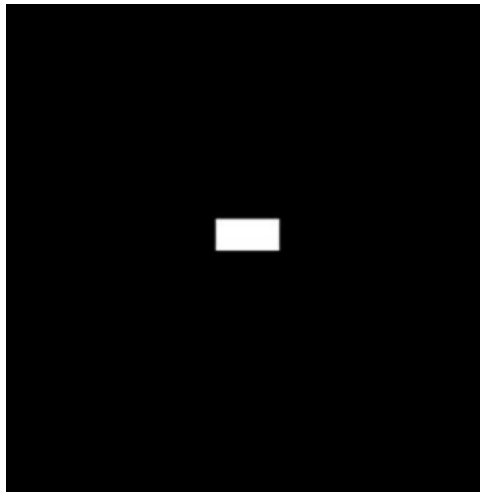
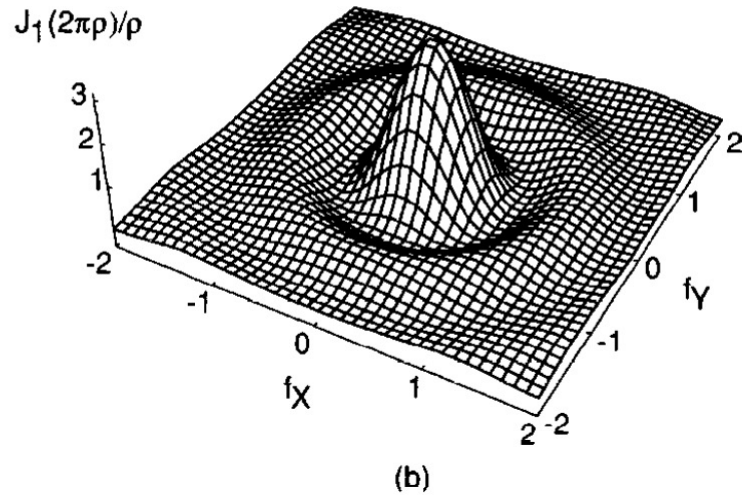
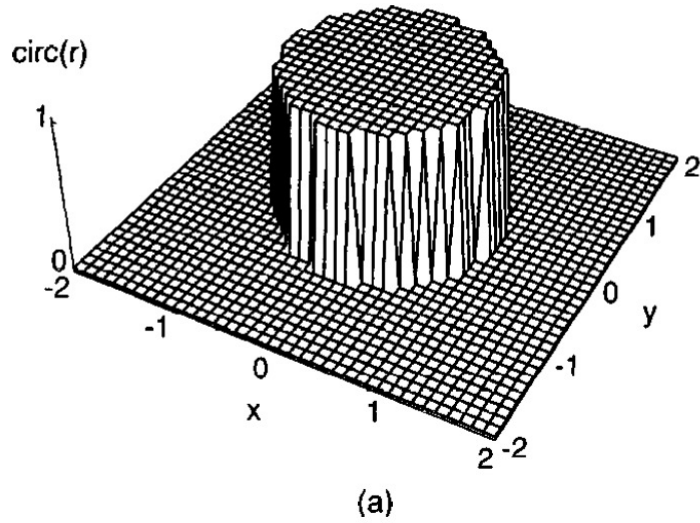
U is absolutely integrable & no infinite discontinuities. The inverse Fourier transform is,

$$\mathcal{F}^{-1}\{\hat{U}(f_x, f_y)\} = U(x, y) = \iint_{-\infty}^{\infty} \hat{U}(f_x, f_y) \exp(2\pi i(f_x x + f_y y)) df_x df_y$$

Additional Details:

- Goodman Chapter 2.1
- Mathworld/Wikipedia, Fourier Transform

Examples of Fourier transform pairs, 2D



$$U_1(x,y)$$



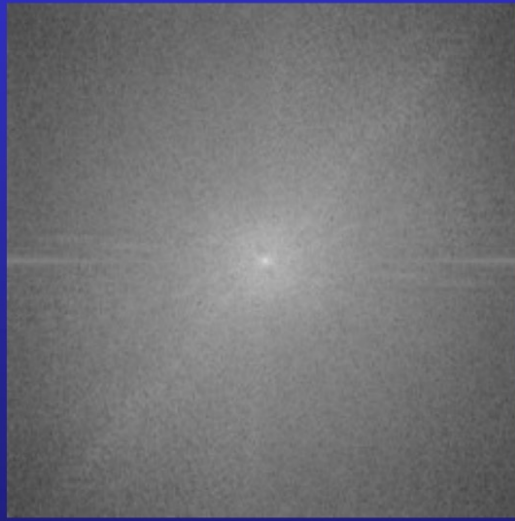
Cheetah

$$U_2(x,y)$$

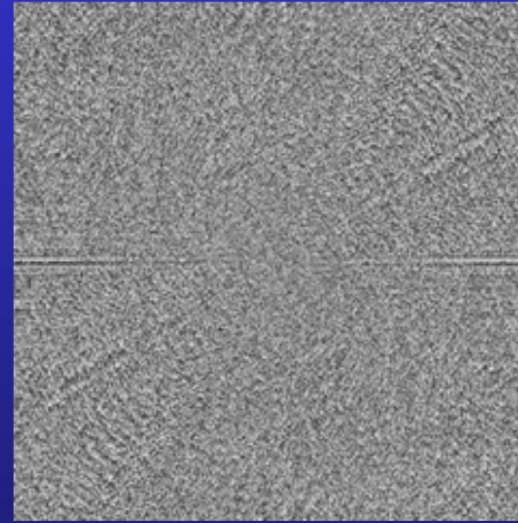


Zebra

$$\hat{U}_1(f_x, f_y)$$

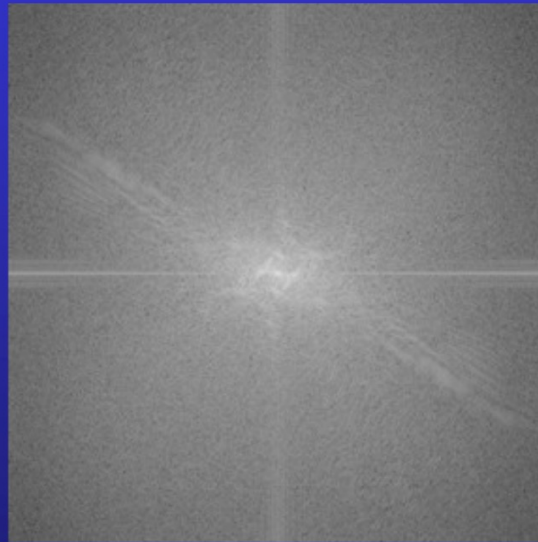


magnitude of cheetah

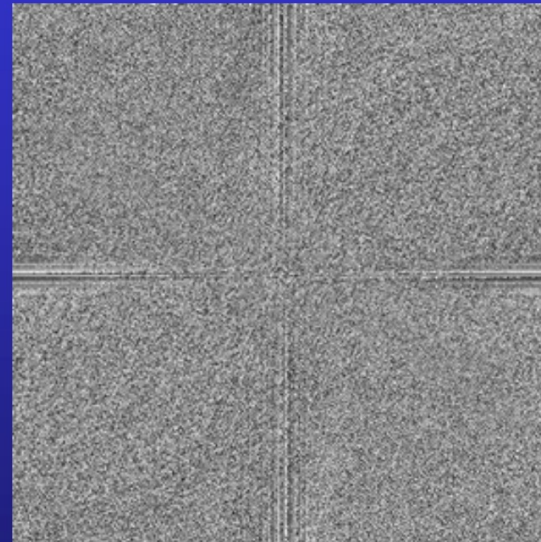


phase of cheetah

$$\hat{U}_2(f_x, f_y)$$



magnitude of zebra



phase of zebra

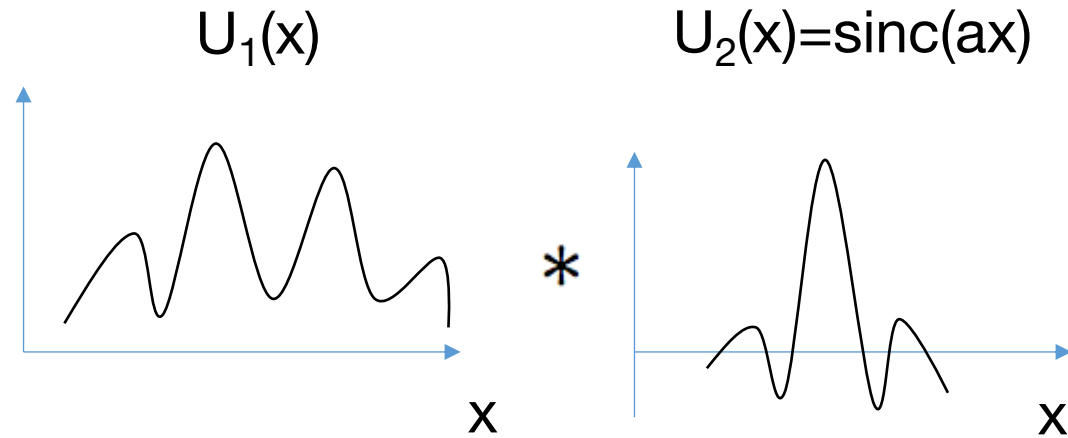
Convolution - Fourier Transform relationship: Convolution Theorem

Convolution theorem. If $\mathcal{F}\{g(x, y)\} = G(f_X, f_Y)$ and $\mathcal{F}\{h(x, y)\} = H(f_X, f_Y)$, then

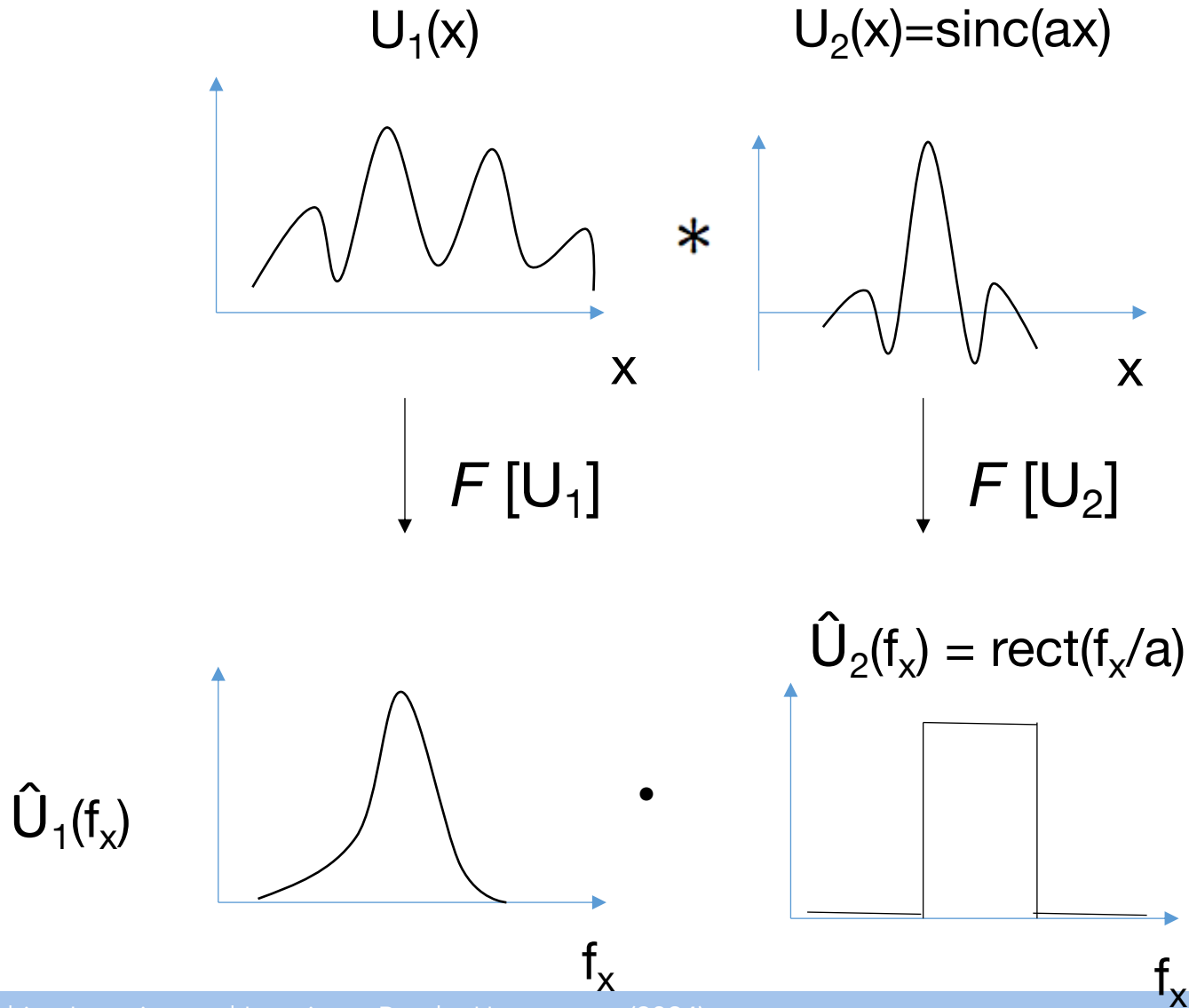
$$\mathcal{F}\left\{\iint_{-\infty}^{\infty} g(\xi, \eta) h(x - \xi, y - \eta) d\xi d\eta\right\} = G(f_X, f_Y) H(f_X, f_Y).$$

“The convolution of two functions in space can be performed by a multiplication in the Fourier domain (spatial frequency domain)”

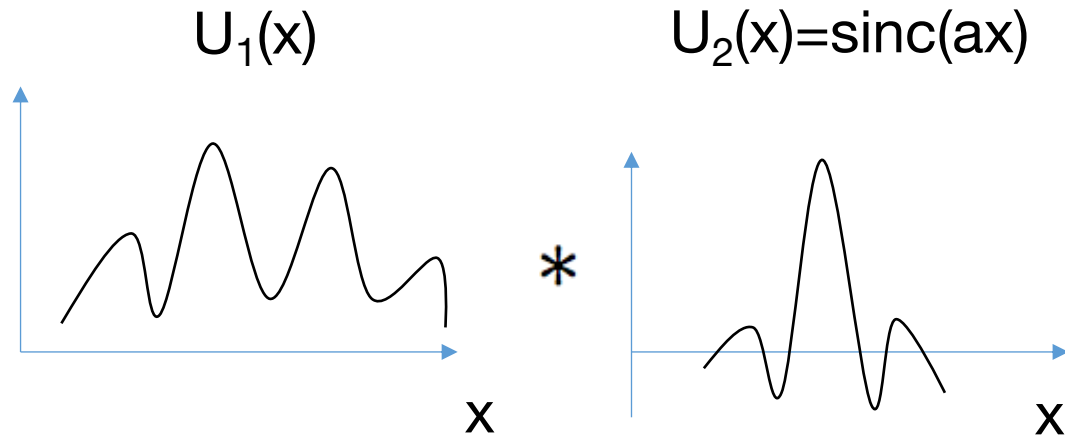
Example of convolution theorem, 1D



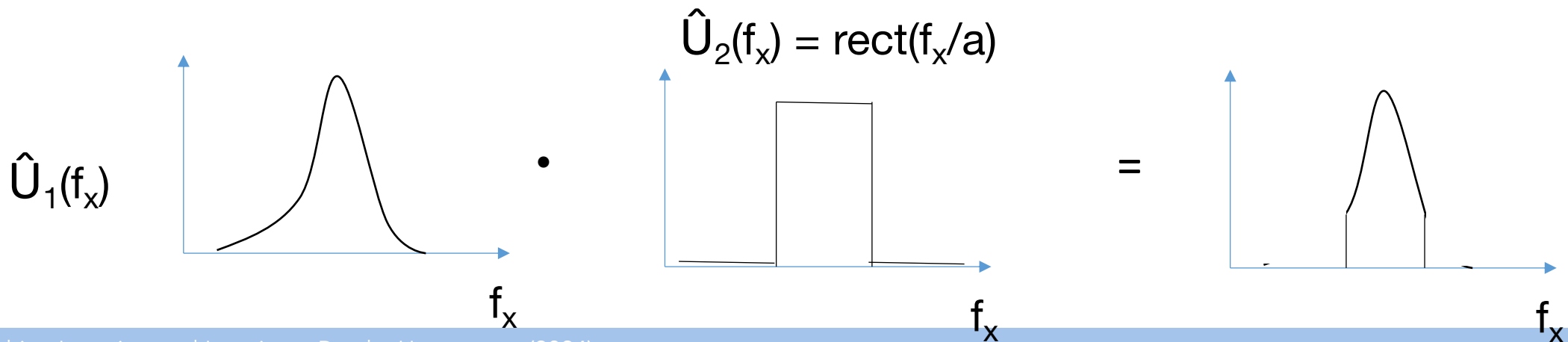
Example of convolution theorem, 1D



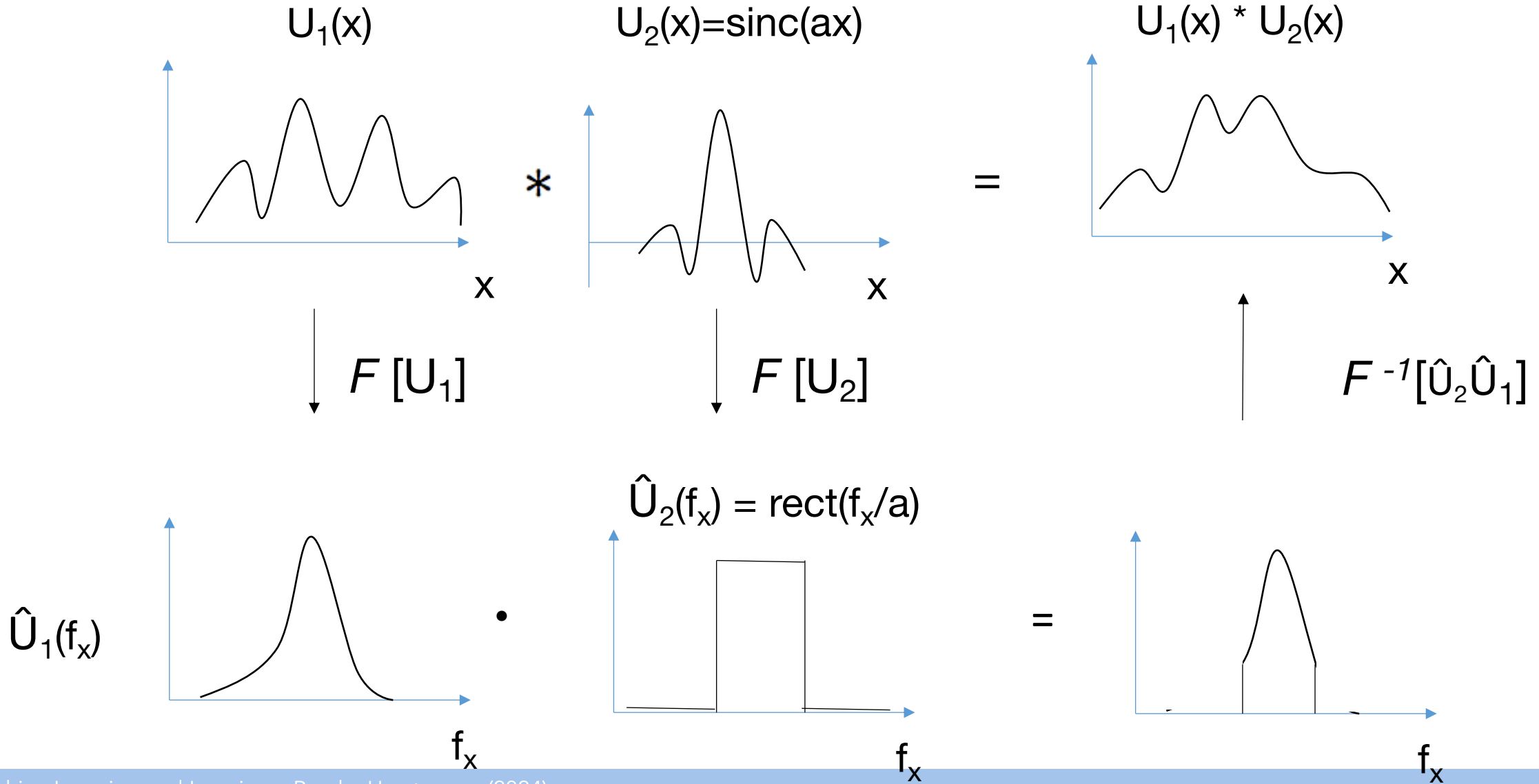
Example of convolution theorem, 1D



$\downarrow F[U_1]$ $\downarrow F[U_2]$

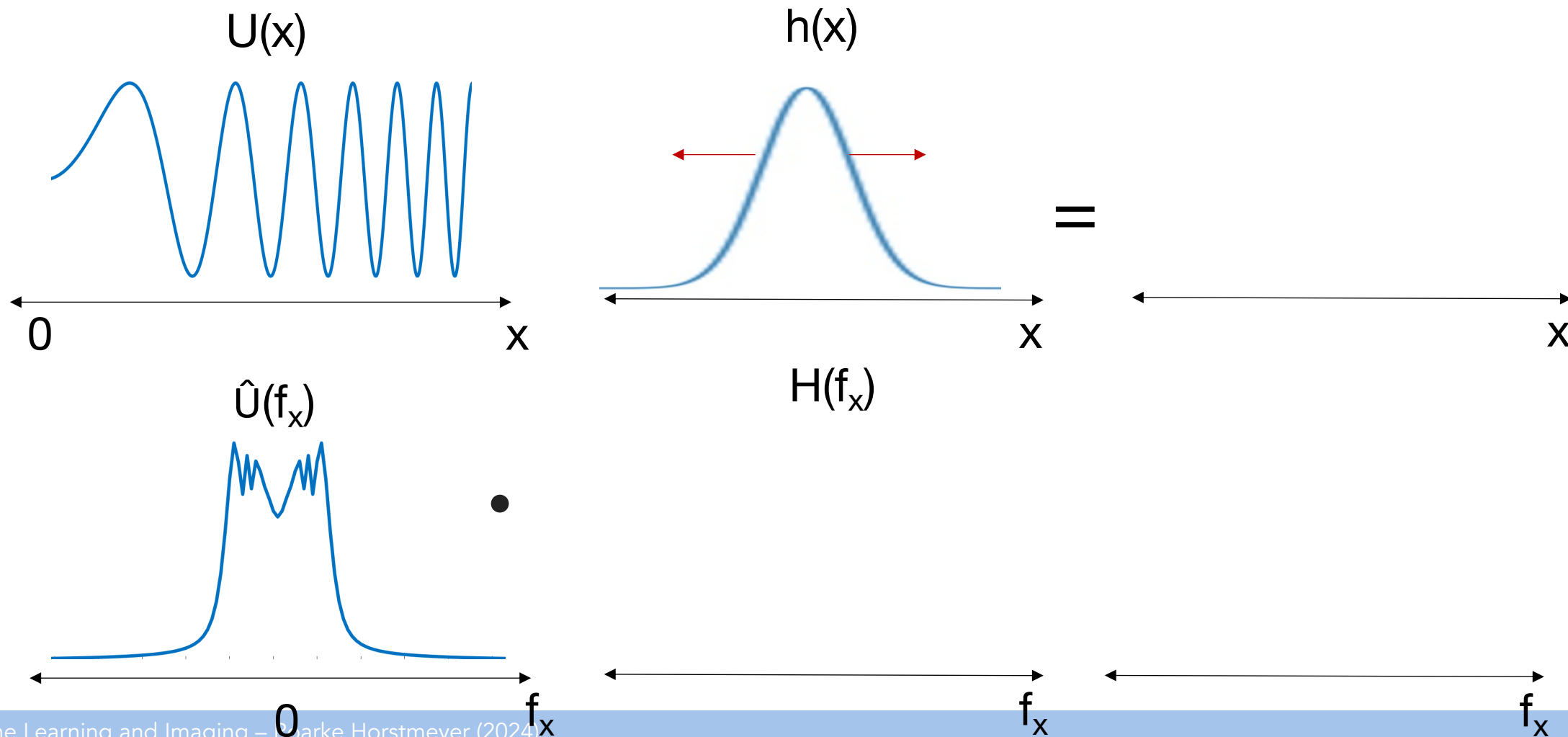


Example of convolution theorem, 1D



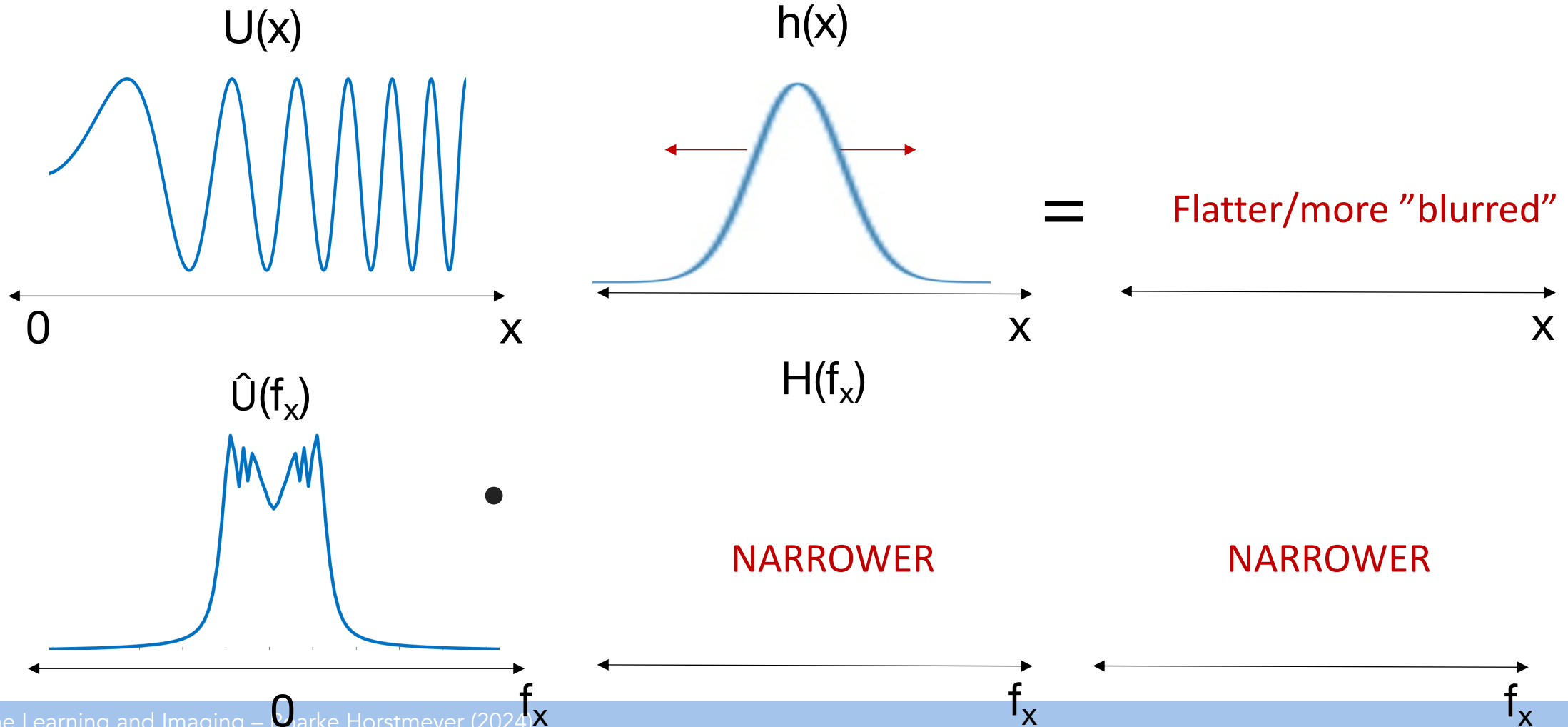
Conceptual questions:

2. Repeat with wider convolution filter:

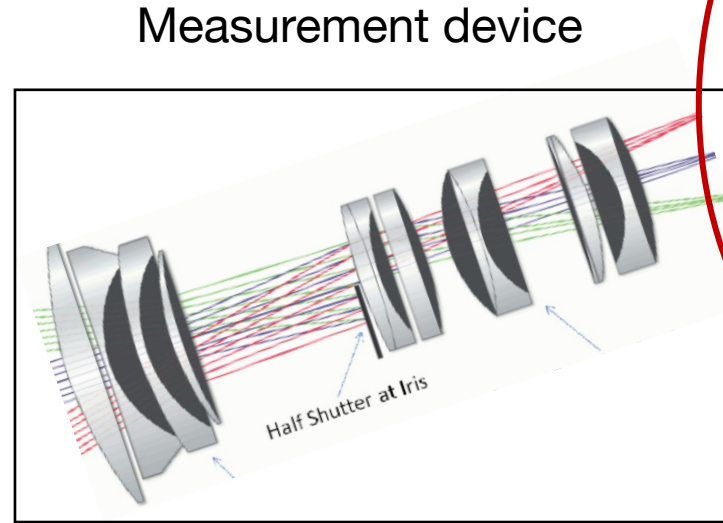
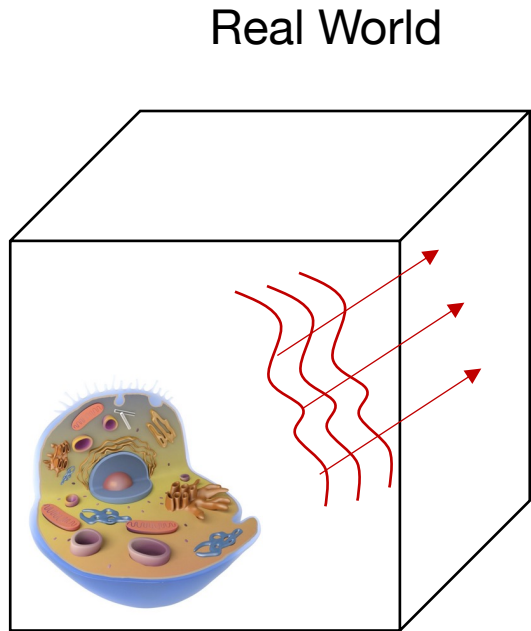
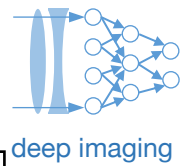


Conceptual questions:

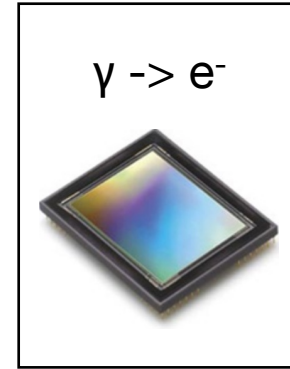
2. Repeat with wider convolution filter:



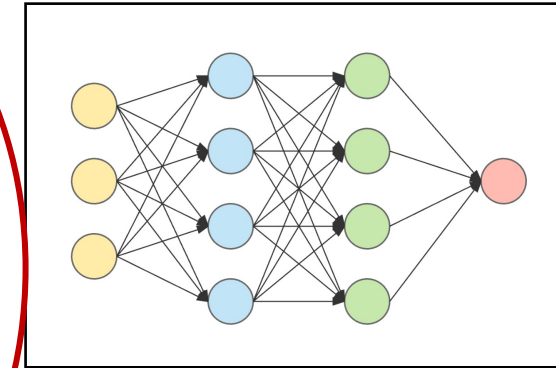
ML+Imaging pipeline introduction



Digitization

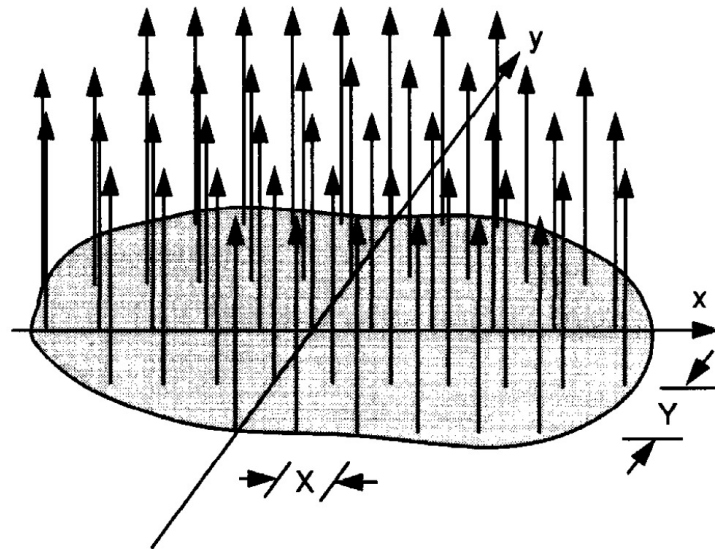


Machine Learning



The Sampling Theorem – from Goodman Section 2.4.1

$$U_s(x, y) = \text{comb}(x/X)\text{comb}(y/Y)U(x, y)$$



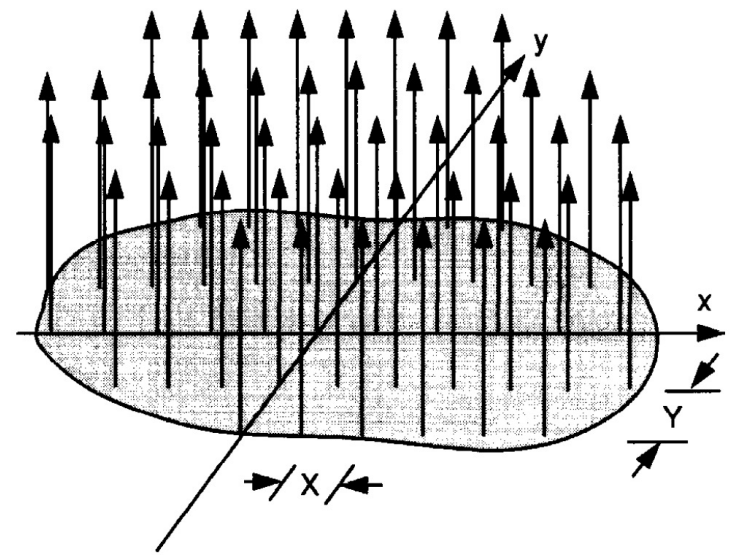
Signal sampling occurs with:

- CMOS (pixel) sensors, PMTs, SPADs
- A-to-D after antennas
- A-to-D after acoustic transducers

Sampling interval width X and Y

The Sampling Theorem – from Goodman Section 2.4.1

$$U_s(x, y) = \text{comb}(x/X)\text{comb}(y/Y)U(x, y)$$



Signal sampling occurs with:

- CMOS (pixel) sensors, PMTs, SPADs
- A-to-D after antennas
- A-to-D after acoustic transducers

Sampling interval width X and Y

$$\hat{U}_s(f_x, f_y) = \mathcal{F} [\text{comb}(x/X)\text{comb}(y/Y)] * \hat{U}(f_x, f_y)$$

The Sampling Theorem – from Goodman Section 2.4.1

$$\hat{U}_s(f_x, f_y) = \mathcal{F} [\text{comb}(x/X)\text{comb}(y/Y)] * \hat{U}(f_x, f_y)$$

The Sampling Theorem – from Goodman Section 2.4.1

$$\hat{U}_s(f_x, f_y) = \mathcal{F} [\text{comb}(x/X)\text{comb}(y/Y)] * \hat{U}(f_x, f_y)$$

$$\mathcal{F} [\text{comb}(x/X)\text{comb}(y/Y)] = \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} \delta \left(f_x - \frac{n}{X}, f_y - \frac{m}{Y} \right)$$

The Sampling Theorem – from Goodman Section 2.4.1

$$\hat{U}_s(f_x, f_y) = \mathcal{F} [\text{comb}(x/X)\text{comb}(y/Y)] * \hat{U}(f_x, f_y)$$

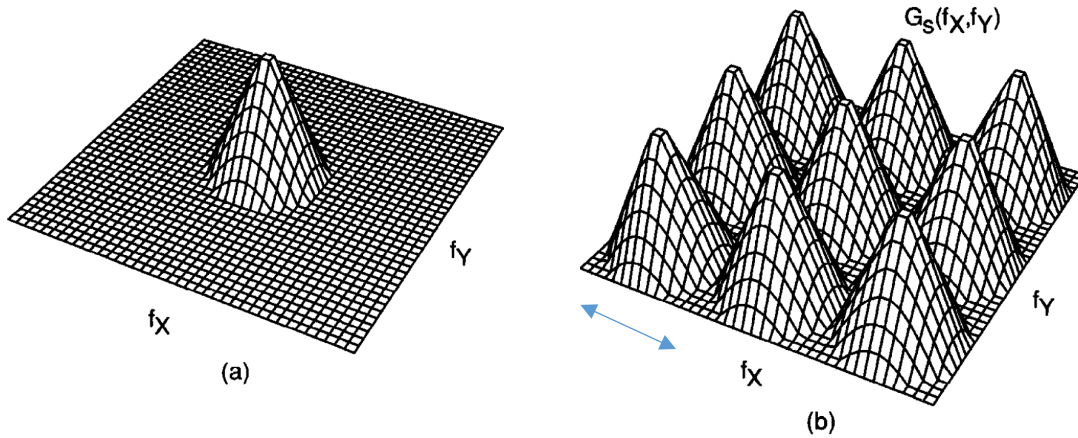
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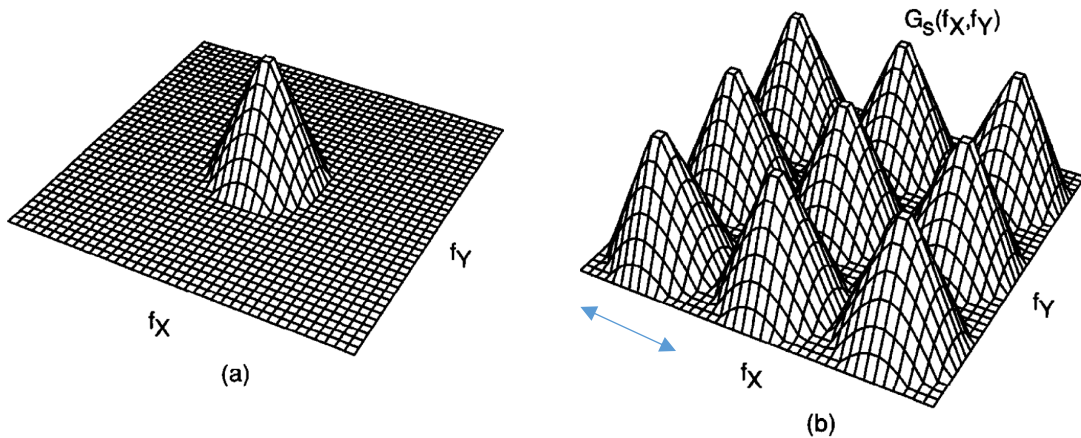


Signal extends from $(-B_x, -B_y)$ to (B_x, B_y) in Fourier domain

The Sampling Theorem – from Goodman Section 2.4.1

$$\mathcal{F} [\text{comb}(x/X)\text{comb}(y/Y)] = \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} \delta \left(f_x - \frac{n}{X}, f_y - \frac{m}{Y} \right)$$

$$\hat{U}_s(f_x, f_y) = \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} \hat{U} \left(f_x - \frac{n}{X}, f_y - \frac{m}{Y} \right)$$



Signal extends from $(-B_x, -B_y)$ to (B_x, B_y) in Fourier domain

Mask out copies with a rect function:

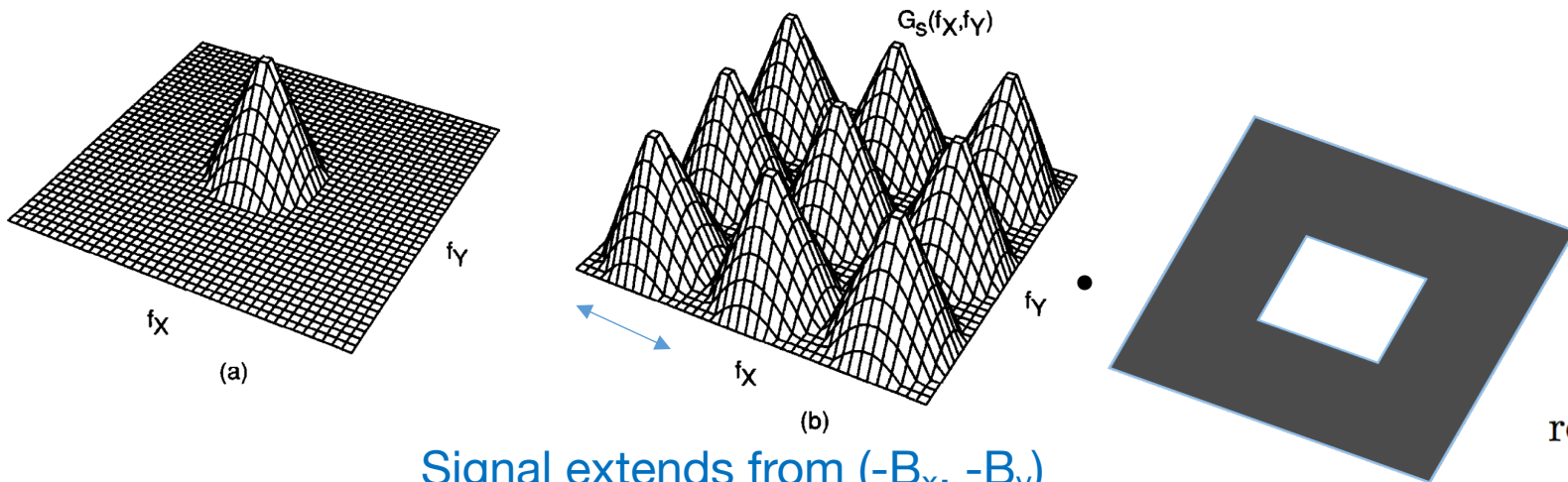
$$\text{rect} \left(\frac{f_x}{2B_x} \right) \text{rect} \left(\frac{f_y}{2B_y} \right) \hat{U}_s(f_x, f_y) = \hat{U}(f_x, f_y)$$

Bandwidth (B_x, B_y) of signal

The Sampling Theorem – from Goodman Section 2.4.1

$$\mathcal{F} [\text{comb}(x/X)\text{comb}(y/Y)] = \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} \delta \left(f_x - \frac{n}{X}, f_y - \frac{m}{Y} \right)$$

$$\hat{U}_s(f_x, f_y) = \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} \hat{U} \left(f_x - \frac{n}{X}, f_y - \frac{m}{Y} \right)$$



Signal extends from $(-B_x, -B_y)$ to (B_x, B_y) in Fourier domain

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\downarrow $F[\bullet]$ \downarrow $F[\bullet]$

$$\text{rect}\left(\frac{f_x}{2B_x}\right)\text{rect}\left(\frac{f_y}{2B_y}\right)\hat{U}_s(f_x, f_y) = \hat{U}(f_x, f_y)$$

$F[\bullet]$ ↓

$$U(x, y)\text{comb}(x/X)\text{comb}(y/Y)$$

$$\text{rect}\left(\frac{f_x}{2B_x}\right)\text{rect}\left(\frac{f_y}{2B_y}\right)\hat{U}_s(f_x, f_y) = \hat{U}(f_x, f_y)$$

$$F[\bullet] \rightarrow h(x, y) = 4B_x B_y \text{sinc}(2B_x x) \text{sinc}(2B_y y)$$

$$\text{rect}\left(\frac{f_x}{2B_x}\right)\text{rect}\left(\frac{f_y}{2B_y}\right)\hat{U}_s(f_x, f_y) = \hat{U}(f_x, f_y)$$

$$F[\bullet] \rightarrow h(x, y) = 4B_x B_y \text{sinc}(2B_x x) \text{sinc}(2B_y y)$$

$$h(x, y) * (U(x, y) \text{comb}(x/X) \text{comb}(y/Y)) = U(x, y)$$

$$\text{rect}\left(\frac{f_x}{2B_x}\right)\text{rect}\left(\frac{f_y}{2B_y}\right)\hat{U}_s(f_x, f_y) = \hat{U}(f_x, f_y)$$

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$$h(x, y) * (U(x, y) \text{comb}(x/X) \text{comb}(y/Y)) = U(x, y)$$

$$U(x, y) \text{comb}(x/X) \text{comb}(y/Y) = XY \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} U(nX, mY) \delta(x - nX, y - mY)$$

$$\text{rect} \left(\frac{f_x}{2B_x} \right) \text{rect} \left(\frac{f_y}{2B_y} \right) \hat{U}_s(f_x, f_y) = \hat{U}(f_x, f_y)$$

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$$h(x, y) * (U(x, y) \text{comb}(x/X) \text{comb}(y/Y)) = U(x, y)$$

$U_s(x, y)$ (from beginning) =

$$U(x, y) \text{comb}(x/X) \text{comb}(y/Y) = XY \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} U(nX, mY) \delta(x - nX, y - mY)$$

$$U(x, y) = 4B_x B_y XY \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} U(nX, mY) \text{sinc} [2B_x(x - nX)] \text{sinc} [2B_y(y - mY)]$$

The Sampling Theorem

When sampled appropriately, a discrete signal can *exactly* reproduce a continuous signal:

$$\underline{U(x, y)} = 4B_x B_y XY \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} \underline{U(nX, mY)} \text{sinc} [2B_x(x - nX)] \text{sinc} [2B_y(y - mY)]$$

Continuous signal:

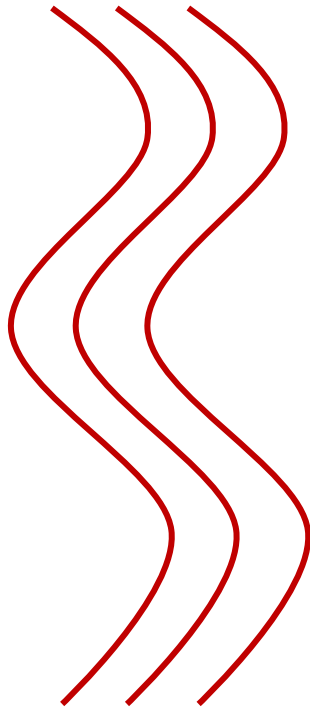
- EM field
- Sound wave
- MR signal

Discretized signal:

- Detected EM field
- Sampled sound wave
- Sampled MR signal

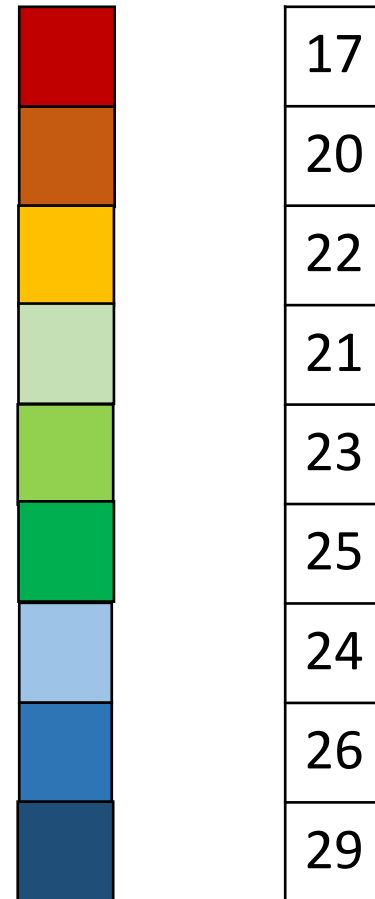
What does the Sampling Theorem mean for us?

Continuous fields



(*) Under certain conditions

Discretize vectors
(and matrices)



Conditions to safely apply the sampling theorem

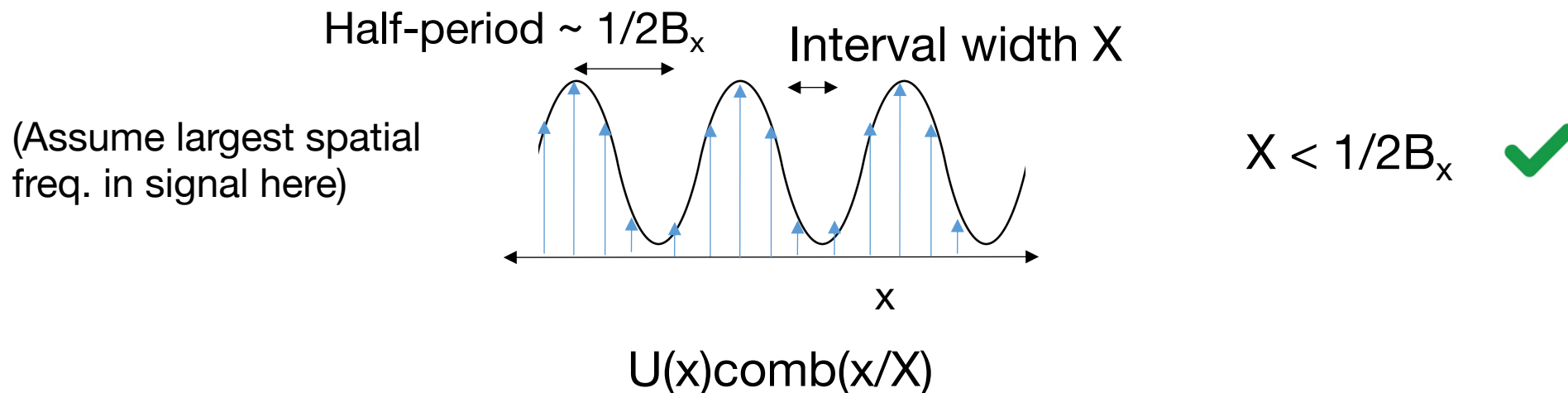
$$U(x, y) = 4B_x B_y XY \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} U(nX, mY) \text{sinc} [2B_x(x - nX)] \text{sinc} [2B_y(y - mY)]$$

- Sampling must be proportional to bandwidth ($2B_x$ and $2B_y$)
 - “Nyquist” sampling: $X = 1/2B_x$, $Y = 1/2B_y$

Conditions to safely apply the sampling theorem

$$U(x, y) = 4B_x B_y XY \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} U(nX, mY) \text{sinc} [2B_x(x - nX)] \text{sinc} [2B_y(y - mY)]$$

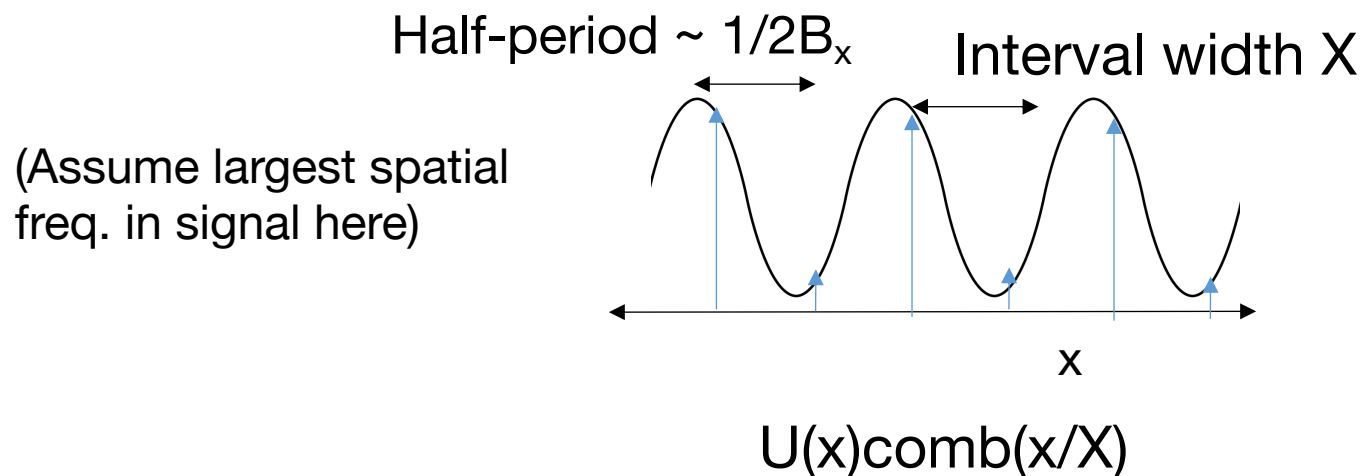
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Conditions to safely apply the sampling theorem

$$U(x, y) = 4B_x B_y XY \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} U(nX, mY) \text{sinc} [2B_x(x - nX)] \text{sinc} [2B_y(y - mY)]$$

- Sampling must be proportional to bandwidth ($2B_x$ and $2B_y$)
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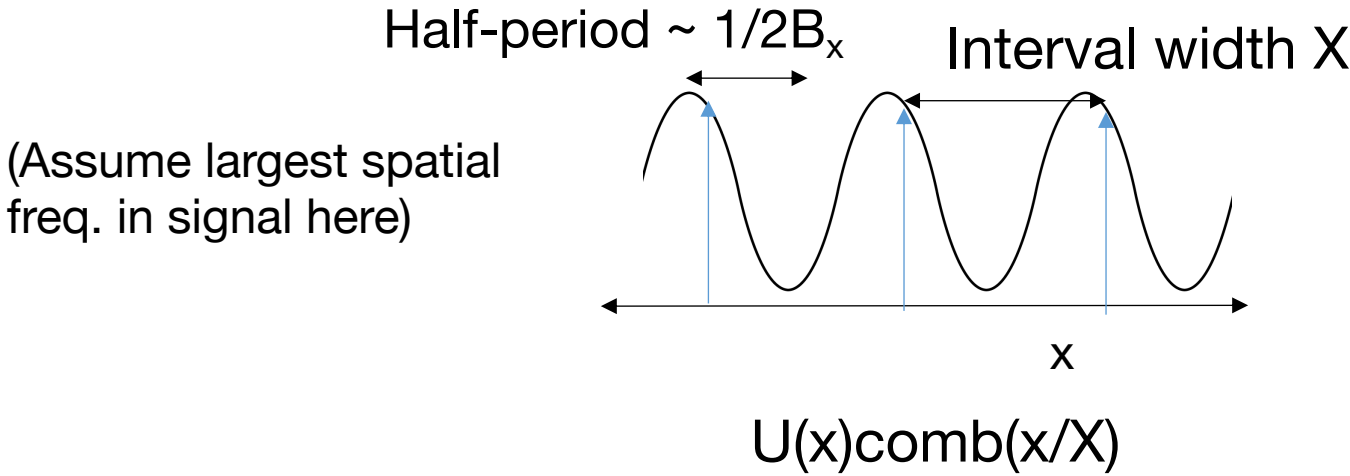
$$X = 1/2B_x \quad \checkmark$$


Nyquist sampling – still sampling peak and trough

Conditions to safely apply the sampling theorem

$$U(x, y) = 4B_x B_y XY \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} U(nX, mY) \text{sinc} [2B_x(x - nX)] \text{sinc} [2B_y(y - mY)]$$

- Sampling must be proportional to bandwidth ($2B_x$ and $2B_y$)
 - “Nyquist” sampling: $X = 1/2B_x$, $Y = 1/2B_y$



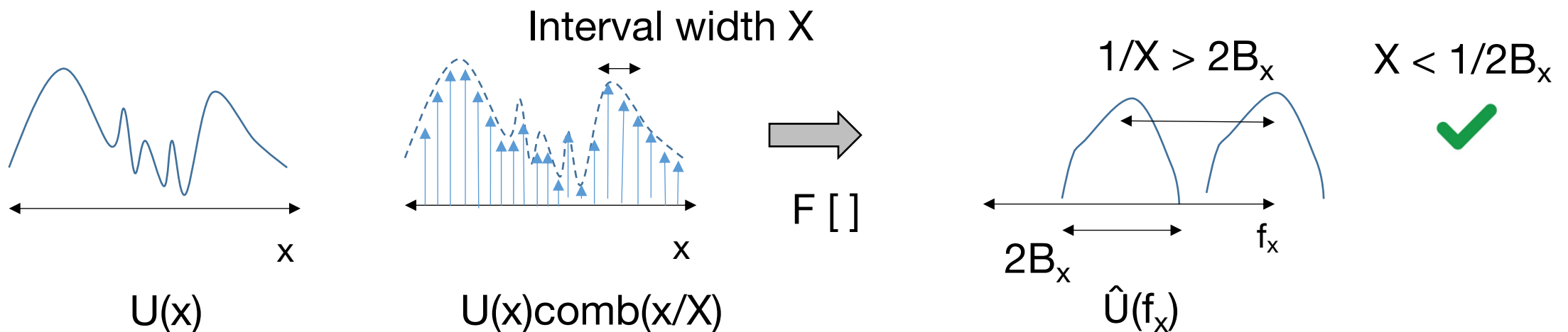
$X > 1/2B_x$ 

Can't detect the frequency anymore!

Conditions to safely apply the sampling theorem

$$U(x, y) = 4B_x B_y XY \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} U(nX, mY) \text{sinc} [2B_x(x - nX)] \text{sinc} [2B_y(y - mY)]$$

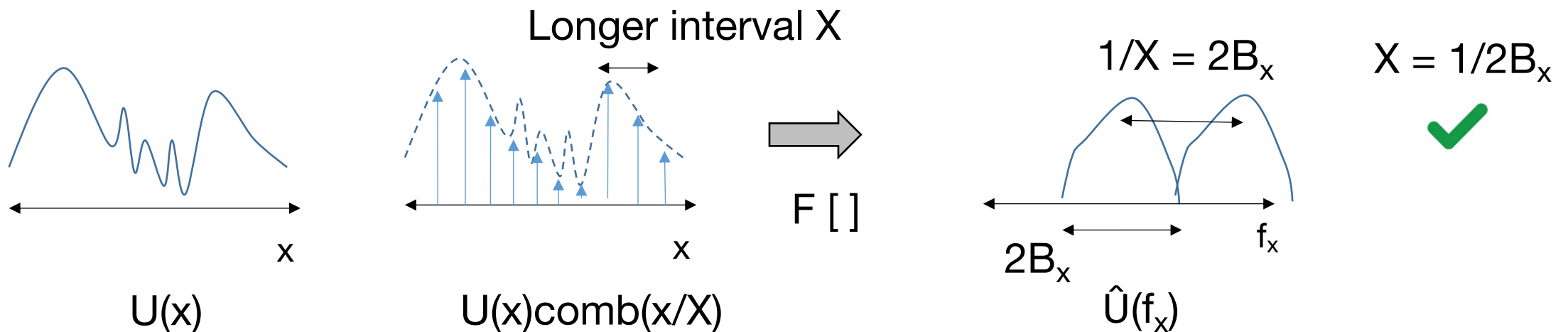
- Sampling must be proportional to bandwidth ($2B_x$ and $2B_y$)
 - “Nyquist” sampling: $X = 1/2B_x$, $Y = 1/2B_y$
 - Needed to avoid aliasing



Conditions to safely apply the sampling theorem

$$U(x, y) = 4B_x B_y XY \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} U(nX, mY) \text{sinc} [2B_x(x - nX)] \text{sinc} [2B_y(y - mY)]$$

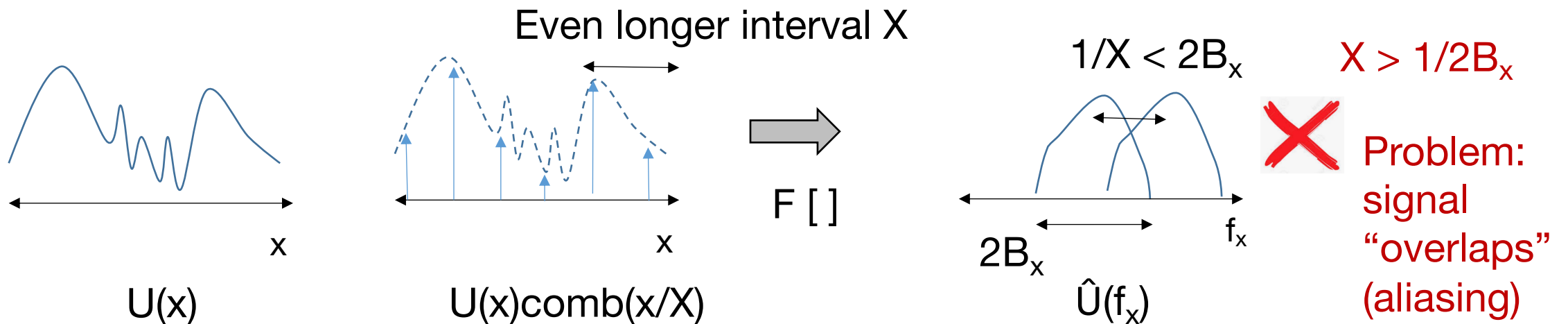
- Sampling must be proportional to bandwidth ($2B_x$ and $2B_y$)
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Conditions to safely apply the sampling theorem

$$U(x, y) = 4B_x B_y XY \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} U(nX, mY) \text{sinc} [2B_x(x - nX)] \text{sinc} [2B_y(y - mY)]$$

- Sampling must be proportional to bandwidth ($2B_x$ and $2B_y$)
 - “Nyquist” sampling: $X = 1/2B_x$, $Y = 1/2B_y$
 - Needed to avoid aliasing



Linear Algebra – notation and basics

- We'll (try to) write *column* vectors as lower case variables
- Row vectors will be denoted as the transpose
- We'll try to write matrices as upper case variables
- We'll try to denote if a matrix/vector is real, complex etc. and its size with a certain notation

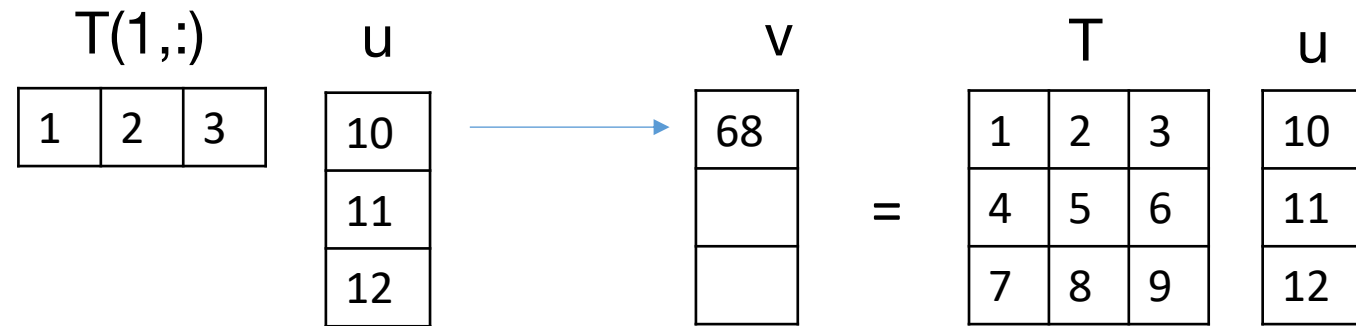
Linear Algebra – notation and basics

Some basic vector operations you should know:

- Conjugate, transpose, conjugate transpose
- Inner product
- Hadamard (element-wise, dot-times) product
- outer product
- Vector (matrix) addition
- matrix-vector product
- convolution

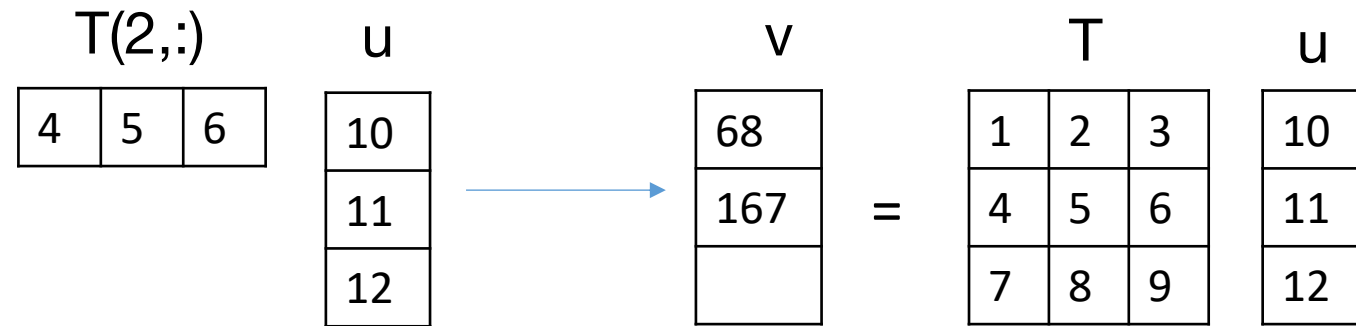
Matrix-vector products – two useful interpretations

1. Inner products per entry:



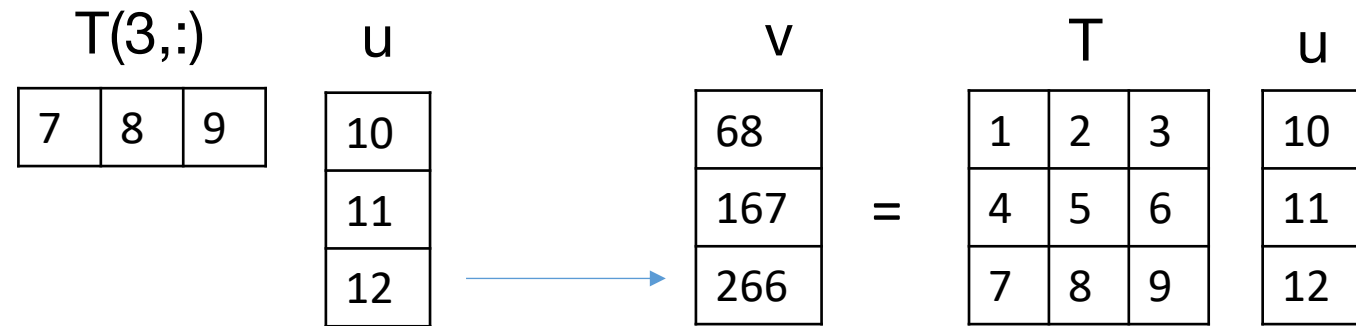
Matrix-vector products – two useful interpretations

1. Inner products per entry:



Matrix-vector products – two useful interpretations

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Matrix-vector products – two useful interpretations

1. Inner products per entry:

$$\begin{array}{c} \mathbf{v} \\ \hline 68 \\ \hline 167 \\ \hline 266 \end{array} = \begin{array}{ccc} & \mathbf{T} & \\ \hline 1 & 2 & 3 \\ \hline 4 & 5 & 6 \\ \hline 7 & 8 & 9 \end{array} \begin{array}{c} \mathbf{u} \\ \hline 10 \\ \hline 11 \\ \hline 12 \end{array}$$

2. Weighted column sum:

$$\begin{array}{c} \mathbf{v} \\ \hline 68 \\ \hline 167 \\ \hline 266 \end{array} = 10 \begin{array}{c} \mathbf{T(:,1)} \\ \hline 1 \\ \hline 4 \\ \hline 7 \end{array} + 11 \begin{array}{c} \mathbf{T(:,2)} \\ \hline 2 \\ \hline 5 \\ \hline 8 \end{array} + 12 \begin{array}{c} \mathbf{T(:,3)} \\ \hline 3 \\ \hline 6 \\ \hline 9 \end{array}$$

Discrete convolution

$$V(x_o) = \int_{-\infty}^{\infty} U(x_i)h(x_o - x_i)dx_i$$



$$v[x_0] = \sum_{x_i=-M}^M u[x_i]h[x_o - x_i]$$

Discrete 1D Convolution – an example

Steps to follow:

Step 1	List the index 'k' covering a sufficient range
Step 2	List the input $x[k]$
Step 3	Obtain the reversed sequence $h[-k]$, and align the rightmost element of $h[n-k]$ to the leftmost element of $x[k]$
Step 4	Cross-multiply and sum the nonzero overlap terms to produce $y[n]$
Step 5	Slide $h[n-k]$ to the right by one position
Step 6	Repeat step 4; stop if all the output values are zero or if required.

<http://host.uniroma3.it/laboratori/sp4te/teaching/sp4bme/documents/LectureConvolution.pdf>

Example 2: Find the convolution of the two sequences $x[n]$ and $h[n]$ given by,

$$x[k] = [3 \ 1 \ 2] \quad h[k] = [3 \ 2 \ 1]$$

↑
↑

k:	-2	-1	0	1	2	3	4	5
x[k]:			3	1	2			
h[-k]:	1	2	3					
h[1-k]:		1	2	3				
h[2-k]:			1	2	3			
h[3-k]:				1	2	3		
h[4-k]:					1	2	3	
h[5-k]:						1	2	3

Hint: The value of k starts from **(- length of $h + 1$)** and continues till **(length of $h +$ length of $x - 1$)**

Here k starts from $-3 + 1 = -2$ and continues till $3 + 3 - 1 = 5$

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y: **9**

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y:
9
6+3

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y:
9
6+3
3+2+6

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 \uparrow

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h[3-k]:					1	2	3				
h[4-k]:							1	2	3		
h[5-k]:									1	2	3

y:
9
6+3
3+2+6
1+4+0

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h[5-k]:							1	2	3

$y:$
9
6+3
3+2+6
1+4+0

$y:$
[9
9
11
5
2
0]

Discrete convolution

$$V(x_o) = \int_{-\infty}^{\infty} U(x_i)h(x_o - x_i)dx_i$$



$$v[x_0] = \sum_{x_i=-M}^M u[x_i]h[x_o - x_i]$$

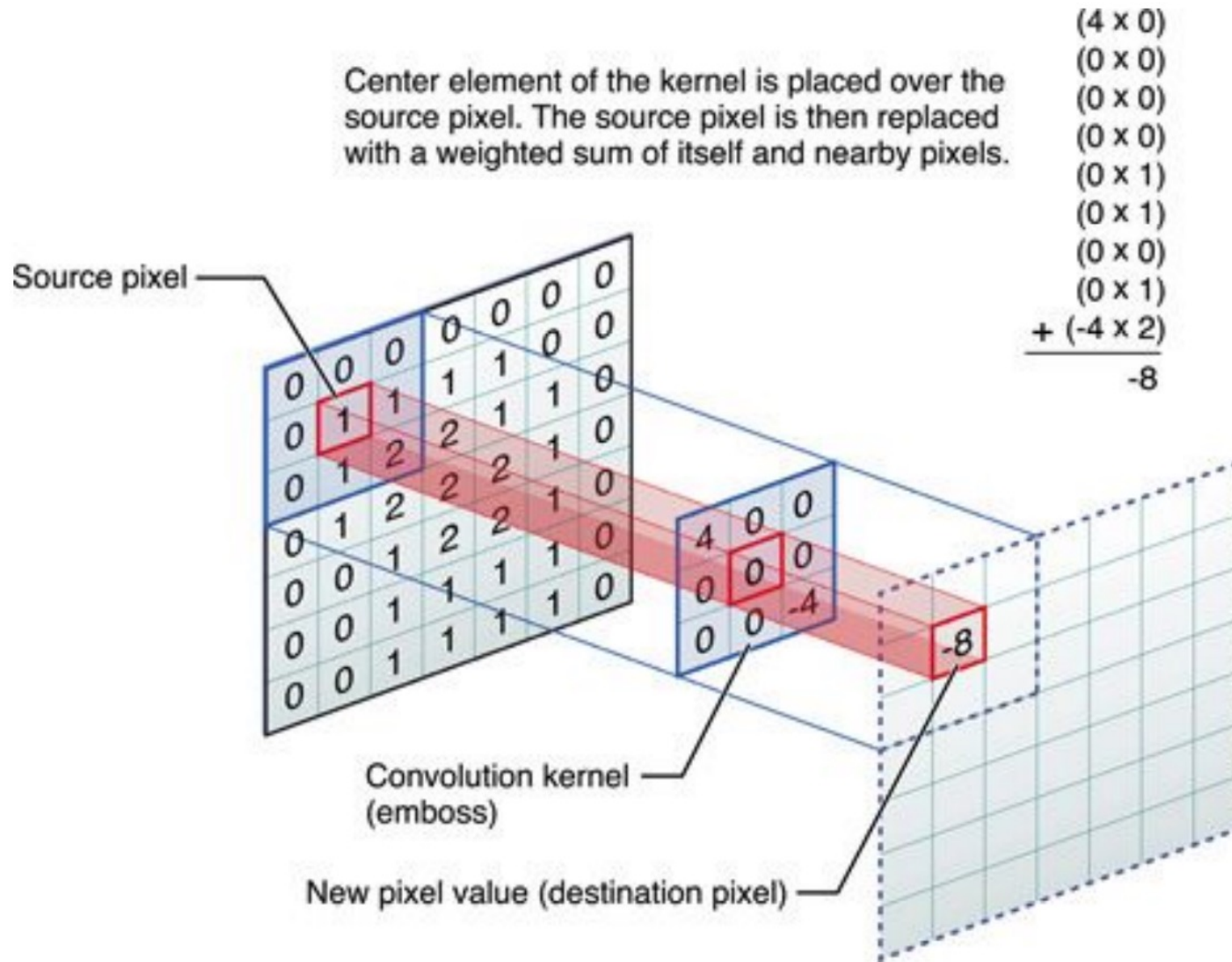
Discrete 2D convolution

$$V(x_o, y_o) = \iint_{-\infty}^{\infty} U(x_i, y_i)h(x_o - x_i, y_o - y_i)dx_idy_i$$



$$v[x_0, y_0] = \sum_{y_i=-L}^L \sum_{x_i=-M}^M u[x_i, y_i]h[x_o - x_i, y_o - y_i]$$

Discrete 2D convolution



<https://www.psi.toronto.edu/~jimmy/ece521/Tut1.pdf>

Discrete 2D convolution: edge conditions and even kernels

From MATLAB definition of conv2:

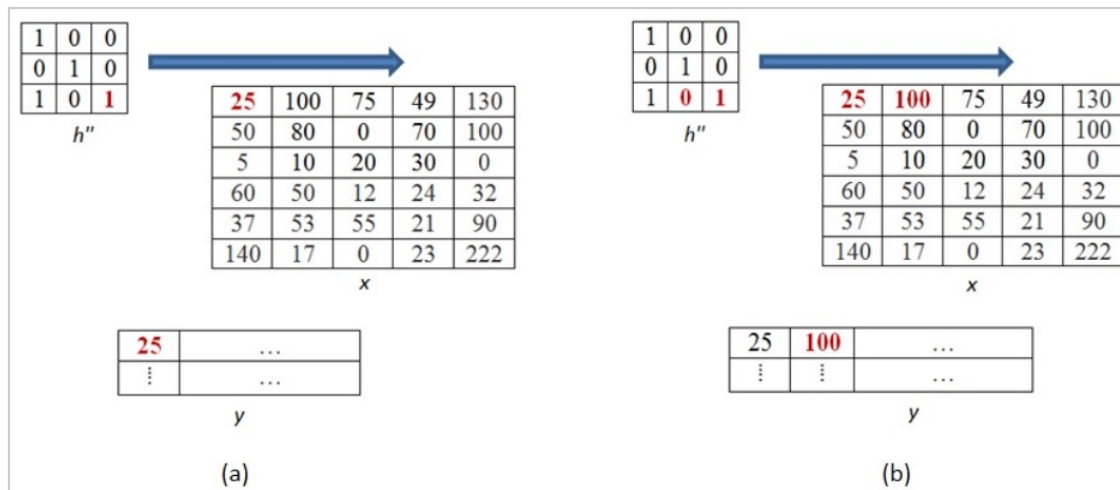
2-D Convolution

For discrete, two-dimensional variables A and B , the following equation defines the convolution of A and B :

$$C(j, k) = \sum_p \sum_q A(p, q) B(j - p + 1, k - q + 1)$$

p and q run over all values that lead to legal subscripts of $A(p, q)$ and $B(j-p+1, k-q+1)$.

$i=1, j=1$: Start in the upper left corner at $A(1,1)$ with *lower right* of flipped version of B [$B(1,1)$]:



Output: 8 x 8

25	100	100	149	205	49	130
50	105	150	225	149	200	100
5	60	130	140	165	179	130
60	55	132	174	74	94	132
37	113	147	96	189	83	90
140	54	253	145	255	137	254
0	140	54	53	78	243	90
0	0	140	17	0	23	222

- For corner-to-corner alignment, doesn't matter if matrix size is even or odd
- Output matrix will be larger than input matrices

Discrete 2D convolution: edge conditions and even kernels

From Tensorflow definition of conv2:

```
output[b, i, j, k] =
  sum_{di, dj, q} input[b, strides[1] * i + di, strides[2] * j + dj, q]
    * filter[di, dj, q, k]
```

1 _{x1}	1 _{x0}	1 _{x1}	0	0
0 _{x0}	1 _{x1}	1 _{x0}	1	0
0 _{x1}	0 _{x0}	1 _{x1}	1	1
0	0	1	1	0
0	1	1	0	0

Image

4		

Convolved Feature

Start convolution kernel *inside* image: align upper-left of image A with *upper right* of kernel B

- Output matrix will be smaller than input image and filter
- We will work through these numbers carefully!

Linear Algebra – notation and basics

Some basic types of matrices & terms that you should know about:

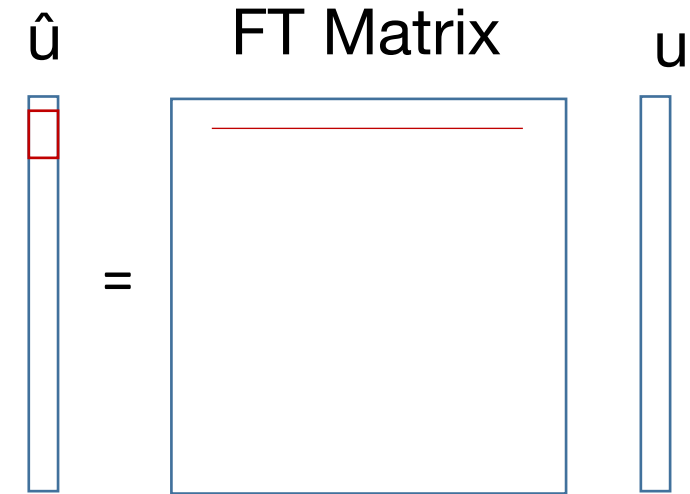
- Symmetric (Hermitian) matrix: $A=A^T$ if A is real, $A=A^H$ if A is complex
- Square, hot-dog and hamburger matrices
- Invertible matrix
- Diagonal matrix
- Toeplitz matrix
- Banded matrix

Discrete Fourier Transforms

$$\hat{U}(f_x) = \int_{-\infty}^{\infty} U(x) \exp(-2\pi i(f_x x)) dx \quad f_x=0$$

$$\hat{u}[f_x] = \sum_{x=0}^{M-1} u[x] \exp(-2\pi i f_x x / M)$$

Inner product of u with different complex expon.



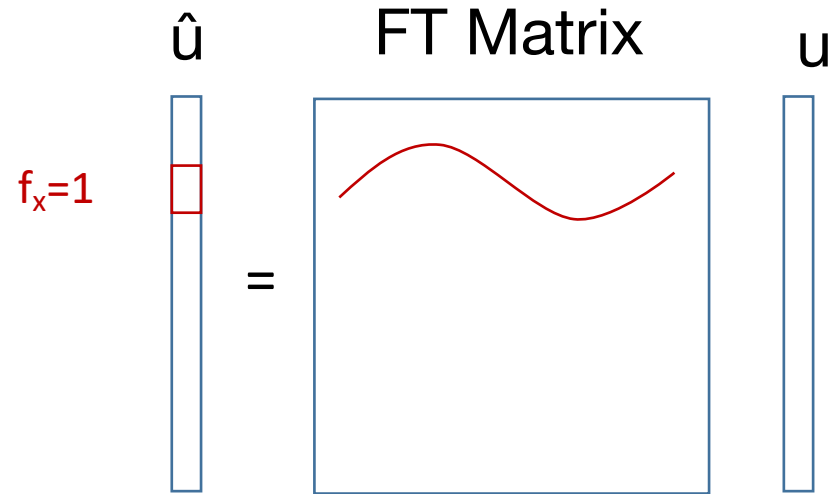
- `np.fft(u)`, `np.fftshift(np.fft(np.ifftshift(u)))`
- `fft` = fast Fourier transform, much more comp. efficient than matrix multiplication!

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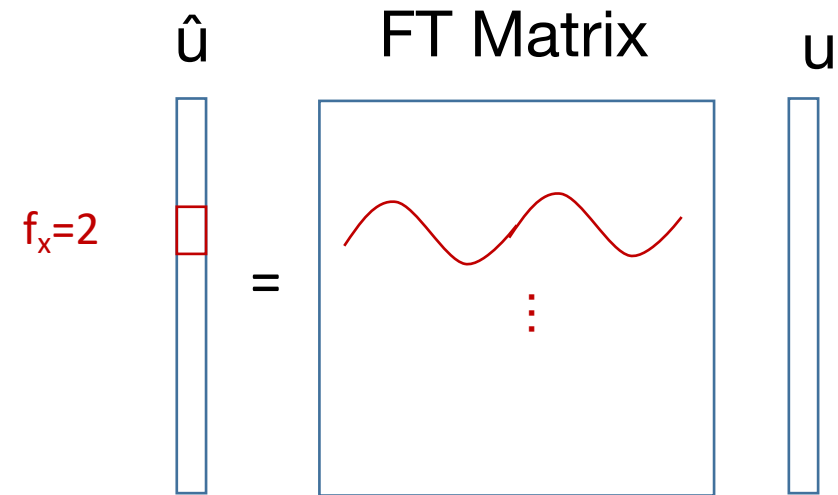
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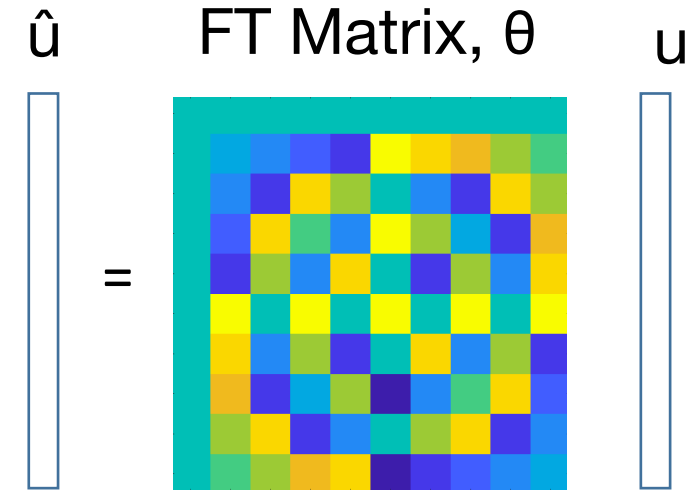


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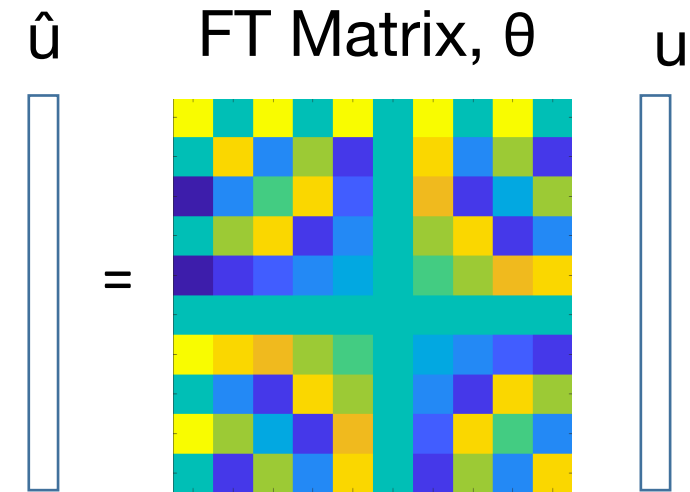
`np.fft(np.eye(10))`

Treats 1st entry of \hat{u} as $f_x=0$

Discrete Fourier Transforms

$$\hat{U}(f_x) = \int_{-\infty}^{\infty} U(x) \exp(-2\pi i(f_x x)) dx$$

$$\hat{u}[f_x] = \sum_{x=0}^{M-1} u[x] \exp(-2\pi i f_x x / M)$$



```
np.fftshift(np.fft(np.ifftshift(np.eye(10))))
```

Treats middle entry of \hat{u} as $f_x=0$

Discrete convolution theorem

Convolution theorem. If $\mathcal{F}\{g(x, y)\} = G(f_x, f_y)$ and $\mathcal{F}\{h(x, y)\} = H(f_x, f_y)$, then

$$\mathcal{F}\left\{\iint_{-\infty}^{\infty} g(\xi, \eta) h(x - \xi, y - \eta) d\xi d\eta\right\} = G(f_x, f_y) H(f_x, f_y). \quad (2-15)$$

Discrete convolution theorem

Convolution theorem. If $\mathcal{F}\{g(x, y)\} = G(f_x, f_y)$ and $\mathcal{F}\{h(x, y)\} = H(f_x, f_y)$, then

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If $\mathcal{F}[g[x, y]] = G[f_x, f_y]$ and $\mathcal{F}[h[x, y]] = H[f_x, f_y]$, and if we know that

$$g[x, y] * h[x, y] = \sum_{l=-L}^L \sum_{m=-M}^M g[m, l] h[x - m, y - l],$$

then from the Convolution Theorem we have,

$$\mathcal{F}[g[x, y] * h[x, y]] = G[f_x, f_y] H[f_x, f_y]$$

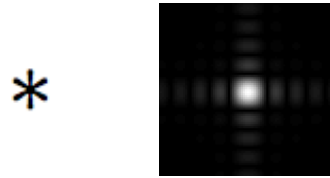
Discrete convolution theorem example –same thing as continuous case

Input image

$U_1(x,y)$



Convolution filter h



=

Output image

$U_2(x,y)$



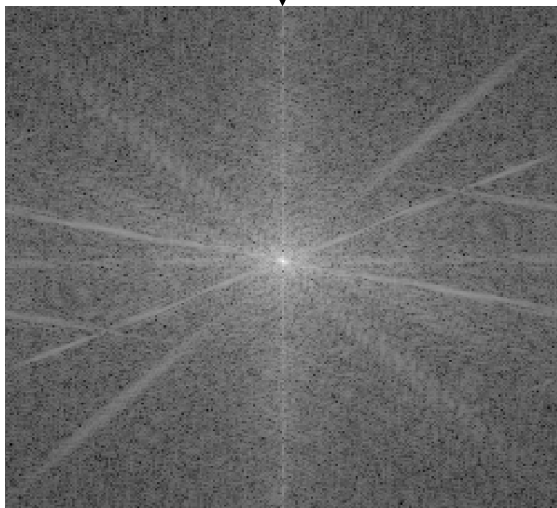
$F [U_1]$

$F [h]$

$F^{-1}[H\hat{U}_1]$

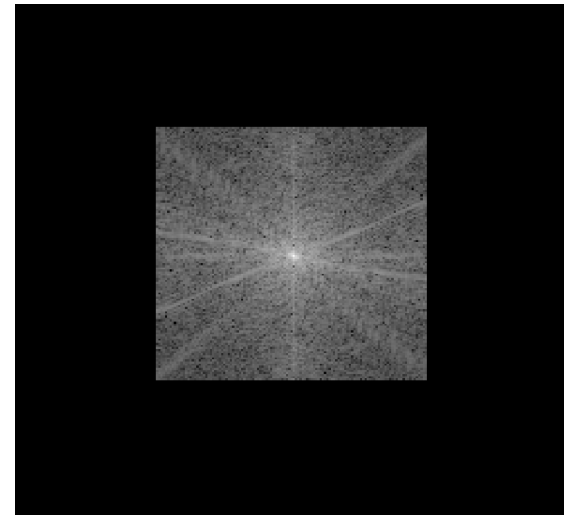
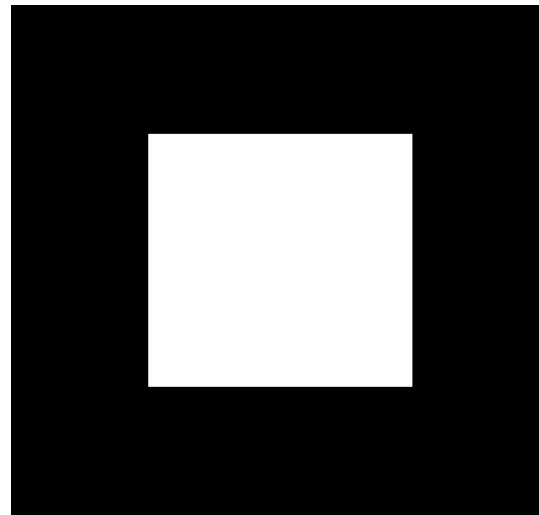
Input spectrum

$\hat{U}_1(f_x, f_y)$



\cdot

=



Convolutions as a big matrix multiplication

$$(u * h)[x] = \sum_{m=0}^{N+M-2} u[m]h[x - m] \longrightarrow y = u * h = \begin{bmatrix} h_1 & 0 & \dots & 0 & 0 \\ h_2 & h_1 & \dots & \vdots & \vdots \\ h_3 & h_2 & \dots & 0 & 0 \\ \vdots & h_3 & \dots & h_1 & 0 \\ h_{m-1} & \vdots & \dots & h_2 & h_1 \\ h_m & h_{m-1} & \vdots & \vdots & h_2 \\ 0 & h_m & \dots & h_{m-2} & \vdots \\ 0 & 0 & \dots & h_{m-1} & h_{m-2} \\ \vdots & \vdots & \vdots & h_m & h_{m-1} \\ 0 & 0 & 0 & \dots & h_m \end{bmatrix} \begin{bmatrix} u_1 \\ u_2 \\ u_3 \\ \vdots \\ u_n \end{bmatrix}$$

Last thing – matrix and vector derivatives

$$\mathbf{u} = \mathbf{W}\mathbf{v}$$

$$\frac{d\mathbf{u}}{d\mathbf{v}} =$$

Last thing – matrix and vector derivatives

$$\mathbf{u} = \mathbf{W}\mathbf{v}$$

$$\frac{d\mathbf{u}}{d\mathbf{v}} =$$

$$\mathbf{u}_3 = W_{3,1}v_1 + W_{3,2}v_2 + \dots + W_{3,M}v_M$$

$$\frac{\partial u_3}{\partial v_2} = \frac{\partial}{\partial v_2}(W_{3,1}v_1 + W_{3,2}v_2 + \dots + W_{3,M}v_M) = \frac{\partial}{\partial v_2}W_{3,2}v_2 = W_{3,2}$$

Last thing – matrix and vector derivatives

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$$\frac{\partial u_i}{\partial v_j} = W_{i,j}$$

$$\frac{d\mathbf{u}}{d\mathbf{v}} = \mathbf{W}$$

Last thing – matrix and vector derivatives

$$\mathbf{u} = \mathbf{W}\mathbf{v}$$

$$\frac{d\mathbf{u}}{d\mathbf{v}} =$$

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$$\frac{\partial u_i}{\partial v_j} = W_{i,j}$$

$$\frac{d\mathbf{u}}{d\mathbf{v}} = \mathbf{W}$$

- When confused, write out one entry, solve derivative and generalize
- Use dimensionality to help (if \mathbf{x} has N elements, and \mathbf{y} has M , then $d\mathbf{y}/d\mathbf{x}$ must be $N \times M$)
- Take advantage of *The Matrix Cookbook*:
 - <https://www.math.uwaterloo.ca/~hwolkowi/matrixcookbook.pdf>